Using Modified Interval Analysis in System Verification¹

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Abstract—Interval arithmetic was original developed to estimate rounding errors in floating-point computations but it is now used in a wide variety of applications from constraint solvers and global optimizers to power and timing analysis of software processes. The objective of interval analysis (IA) is to determine the output ranges (or interval) of a computation set. The main problem of classical interval analysis is the overestimation of the output ranges and its dependency on the coding of the system behavior. In this paper, a modified interval analysis method is presented. The method reduces the interval overestimation and it is independent of the coding. This modified interval analysis is the kernel of a new verification technique that enables the verification of functional properties in system level descriptions and obtains functional test vectors.

Index Terms— Interval Arithmetic, System Level Verification.

I. INTRODUCTION

As system complexity grows, designers describe it at higher abstraction levels and spend more effort on verification. In order to confront this growing complexity, it is essential to define verification methodologies that allow the validation of the design during the specification step, at system level. Simulation is the most widely used verification technique, but even if coverage metrics are used, it has several problems (test bench definition, completeness, etc). Another possibility is to use formal verification techniques. Some of these are based on transforming the system description into a functionally canonical form, such as BDDs [11], and deriving the solution from this structure. Other verification techniques test the satisfiability of properties (e.g. SAT-based verification [2, 7, 8, 9, 10]) or combine both approaches. However, these previously commented approaches generally suffer from exponential worst-case complexity [2], because they use Boolean representations of the system that increase the number of signals and operators during the verification processes.

This paper proposes a methodology that uses a Finite State Machine (FSM) technique to identify execution paths that reach assertions. This technique is combined with a modified non-linear solver to find counterexamples. Our approach assumes that the data path has only arithmetic/relational operations so all Boolean operators are converted into integer arithmetic expressions and control constructions. Thus, the main differences from the previously commented approaches are that the proposed technique uses a higher abstraction level (integer/real equations instead of Boolean equations) and handles multiplications efficiently (non-linear equations). This implies an important cost reduction when designs with a complex data path are verified. The main drawbacks of the proposed approach are that it can not handle efficiently designs with a lot of word-level logic operators or execution paths.

The kernel of the proposed technique is a MODified Interval Analysis (MODIA). The objective of classical IA is to determine the output ranges (or interval) of a computation set [12]. Although this technique has been used previously in verification (e.g. timing and power analysis in software [13]), it has important drawbacks: overestimation of the output range and expression dependency. For example, if the behavior of the 'y' output is represented by the equation 'y=(x*x)-x' and the 'x' input is defined in the range [-2,3], it is possible to derive that the output 'y' will be defined in the range [-9,11] using interval arithmetic [12]. However, if the output is defined by the equivalent equation 'y=x(x-1)', the derived range will be [-9,6]. Both approaches are overestimations of the correct range, [-1/4,6].

The main objective of the proposed technique is to find an input range that verifies a set of properties and/or constraints, reducing the classical interval analysis overestimation and being independent of the coding [14]. Additionally, the proposed algorithm computes points instead of classical intervals, allowing the reduction of the number of computations.

II. SYSTEM MODELING

In our approach, we assume that the system is described at behavioral level as a set of statements that operate with integer data. The addition, subtraction and multiplication operators are directly supported as well as 'if' control statements. Concerning loop statements, a predefined number of iterations are unrolled in a similar way as the time frame expansion in Automatic Test Pattern Generation (ATPG) and Bounded Model Checking [15]. Boolean operations are converted into integer expressions as Fig. 1 shows.

The left column VHDL description (a) contains a single bit (o1) and word-level (o2) Boolean operations. The first operation is directly converted (b column) into an integer expression but the second conversion needs 24 if statements. These control statements increase the number of execution paths and this reduces the algorithm efficiency.

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Variable a b c [.]	Variable a b c: integer range 0 to 1:			
std logic:	Variable m.n. integer range 0 to 1,			
variable m n l· std logic				
vector (7 downto 0):	200,			
	variable (1,t2,t3: Integer range			
	0 to			
	255;			
m.– n or i, oz	a:= b*c; 01			
	t1:=n: t2:=l:t3:=0:			
	if(t1>127 or t2>127) then			
	t3:=t3+128			
	if(t1 > 120) then $t1 = t1 + 120$, $if(t1 > 120)$ and $if(t1 > 120)$			
	II((1/127) then (1(1-120, end II,			
	If(t2>127) then $t2:=t2-128$; end IT;			
	end if; bit 7			
	bit			
	m:=t3; o2			



Other operators (division, etc) are currently supported with the same restrictions as commercial RTL synthesis tools (e.g. Synopsys VHDL Compiler).

With the previously commented consideration, every execution path in the behavioral description can be modeled with polynomials (that model the system behavior) and a set of constraints that model the if-statement conditions and properties/assertions [16].

The main goal of the modified interval analysis (MODIA) is to find the bounds of non-linear polynomial equations. The algorithm is presented in section III, which is divided into five parts. The first part explains the basic MODIA idea for one dimension, the second shows the formal demonstrations of the bounds, the third is an example of interval evaluation, the fourth presents the proposed interval-oriented arithmetic and the last part, the extension to N dimensions. The verification algorithm (that is presented in section IV) uses these bounds to find input integer intervals that satisfy the predefined set of constraints, defined with non-linear polynomial inequalities. Section IV also presents the execution-path extraction algorithm. Finally, section V presents some experimental results.

III. MODIFIED INTERVAL ANALYSIS

A. Basic Idea

At behavioral level, the functionality can be represented by polynomial functions. Let $P(x_1, x_2, ..., x_N)$ be the polynomial that describes the behavior of a system 'S' with an input space of N integer variables that take positive values: $x_1 \in [x_{II}, x_{SI}]$ ($0 \le x_{II} \le x1 \le x_{SI}$), ..., $x_N \in [x_{IN}, x_{SN}]$ ($0 \le x_{IN} \le x_N \le x_{SN}$). The algorithm can also handle systems in which some variables take positive values and others negative values. If the same variable takes positive and negative values, its input interval is split into two parts (positive and negative subinterval) and considered independently by the algorithm. The analysis of intervals with negative values and it has been

omitted to simplify the algorithm.

In order to find a maximum and minimum bound of a polynomial, P(X), its expression is separated into two parts: the positive (P⁺(X)) and negative part (P⁻(X)). A very important property of the positive part (polynomial in which all the monomials take positive values) is that it is an increasing function. Similarly, the negative part (polynomial in which all the monomials take negative values) is a decreasing function. The expression of these polynomials could be very complex (see equation 8) and, in order to bound them, two simple functions are defined: $P_R(X)$ and $P_M(X)$. In the case of a design with only one input, x, the expression of these functions is presented in (1):

$$P_R(x) = mx + c$$

$$P_M(x) = \alpha x^n + \beta$$
(1)

The 'n' parameter is equal to the order of the polynomial that is being bounded (positive or negative part). The 'm', 'c', ' α ' and ' β ' are calculated by evaluating the positive and negative polynomials at the interval extremes (X₁ and X_S). The demonstrates are shown in subsection III.B.

It is easy to see that the original polynomial, P(X), is now bounded by these simple functions (2):

$$P_{R}^{+}(X) \ge P^{+}(X) \ge P_{M}^{+}(X)$$

$$P_{M}^{-}(X) \ge P^{-}(X) \ge P_{R}^{-}(X)$$

$$P(X) = P^{+}(X) + P^{-}(X)$$

$$P_{R}^{+}(X) + P_{M}^{-}(X) \ge P(X) \ge P_{M}^{+}(X) + P_{R}^{-}(X)$$
(2)

The maximum and minimum value of the bound functions (P_R+P_M) can be calculated easily with classical derivativebased algorithms.

B. Formal Demonstrations

Theorem 1. There is a upper bound function $(P_R(x))$ for any increasing polynomial (the positive part) in an interval of input $[x_h, x_s]$.

Proof. Let the positive polynomial, $P^+(x)$, be

$$P^{+}(x) = a_N x^N + a_{N-1} x^{N-1} + \dots + a_1 x + a_N$$

where $a_N > 0$, $a_i \ge 0 \quad \forall i \in [0, N-1]$. In order to simplify the demonstration, the following variable change is carried out. $0 \le t \le 1$

$$x = (x_S - x_I)^* t + x_I$$

The polynomial with the new variable is

$$P^{+}(t) = b_{N}t^{N} + b_{N-1}t^{N-1} + \dots + b_{1}t + b_{0}$$

with $b_N > 0$, $b_i \ge 0 \ \forall i \in [0, N-1]$. The upper bound function $P_R(t) = mt + c$

is the same at the extremities.

$$P^{+}(0) = b_{0} = P_{R}(0) = c$$

$$P^{+}(1) = \sum_{j=0}^{N} b_{j} = P_{R}(1) = m + c$$
(3)

Any t must comply with

$$P_R(t) - P^+(t) \ge 0 \tag{4}$$

Substituting the expressions (3) into (4) and simplifying,

$$P_{R}(t) - P^{+}(t) = \sum_{j=1}^{N} b_{j}t + b_{0} - \sum_{j=0}^{N} b_{j}t^{j} = \sum_{j=1}^{N} b_{j}(t - t^{j}) + (b_{0} - b_{0}) =$$
$$= \sum_{j=1}^{N} b_{j}t(1 - t^{j-1}) \ge 0 \Leftrightarrow t(1 - t^{j-1}) \ge 0 \Leftrightarrow 1 \ge t^{j-1}$$

for all $t \in [0, 1]$.

In the case of the decreasing polynomial (the negative part), the demonstration that $(P_M(x))$ is a lower bound function of P⁻(x), is similar.

Theorem 2. There is a lower bound function $(P_M(x))$ for any increasing positive polynomial in an interval of input $[x_h, x_s]$.

Proof. As the above demonstration, take $P^+(t)$ to simplify the proof.

$$P^{+}(t) = b_{N}t^{N} + b_{N-1}t^{N-1} + \dots + b_{1}t + b_{0}t^{N-1}$$

with $b_N > 0$, $b_i \ge 0 \ \forall i \in [0, N-1]$. The lower bound function $P_M(x) = \alpha x^n + \beta$

is the same at the extremities.

$$P^{+}(0) = b_{0} = P_{M}(0) = \beta$$

$$P^{+}(1) = \sum_{j=0}^{N} b_{j} = P_{M}(1) = \alpha + \beta$$
(5)

(6)

Any t must compy with

$$P^{+}(t) - P_{M}(t) \ge 0$$

Substituting the expressions (5) into (6) and simplifying,

$$P^{+}(t) - P_{M}(t) = \sum_{j=0}^{N} b_{j}t^{j} - \left(\sum_{j=1}^{N} b_{j}t^{n} + b_{0}\right) = \sum_{j=1}^{N} b_{j}(t^{j} - t^{n}) + (b_{0} - b_{0}) =$$
$$= \sum_{j=1}^{N} b_{j}t^{j}(1 - t^{n-j}) \ge 0 \Leftrightarrow t^{j}(1 - t^{n-j}) \ge 0 \Leftrightarrow 1 \ge t^{n-j}$$

for all $t \in [0, 1]$.

In the case of the decreasing polynomial (the negative part), the demonstrations that $(P_R(x))$ is an upper bound function of P(x) is similar.

C. Example

The goal is to bound a one-dimensional polynomial (7). The input space is $x \in [0, 6]$.

$$P(x) = \frac{1}{50} \left(\frac{1}{4} x^6 - 1.9 x^5 + \frac{1}{4} x^4 + 20 x^3 - 17 x^2 - 55.6 x + 48 \right)$$
(7)

As the inputs are always positive, the monomials with positive coefficient constitute the positive part and the rest the negative part. Their equations are shown in (8).

$$\begin{cases} P^{+}(x) = \frac{1}{50} \left(\frac{1}{4} x^{6} + \frac{1}{4} x^{4} + 20x^{3} + 48 \right) \\ P^{-}(x) = \frac{1}{50} \left(-1.9x^{5} - 17x^{2} - 55.6x \right) \end{cases}$$
(8)

In order to calculate the coefficients of P_M and P_R functions, it is only necessary to evaluate the interval extremes (X_I and X_S) (see Table I). In the next section, an efficient algorithm to evaluate these values is presented. In this example, the extreme values are:

x value	$X_{I} = 0$	$X_{s} = 6$			
Positive polynomial	0.96	327.12			
Negative polynomial 0 314.4					
Table I. Values at the extremities of (4)					

Fig. 2 presents the bound of polynomial P(x). Its maximum bound is (9) and its minimum bound is (10).

$$M(x) = -0.044x^5 + 54.36x + 0.96 \tag{9}$$

$$n(x) = 0.00699x^6 - 52.4x + 0.96 \tag{10}$$

In order to calculate the maximum and minimum values, elementary derivative calculus is applied. The result is that the function P(x) is bounded by the interval *[-180.80, 177.06]*. This result overestimates the real range, *[-1.07, 12.7]*, but it is better than the classical IA results, *[-313.44, 327.12]*.

A classical way to improve the results is to split the original interval into two subintervals and apply the algorithm to both. MODIA improves this process thanks to an important property: only the extreme values are needed to calculate the bound functions. Thus, these values can be reused to calculate the subinterval bounds. For example, if we want to calculate the bounds of the [0,3] interval after we have calculated the [0,6] interval, we will only have to evaluate the polynomials P⁺ and P⁻ at the point '3', because the value at the other extremity (point '0') has been calculated during the estimation of the [0,6] interval.

D. Data Flow Graph Evaluation

In order to avoid the extraction of the analytical expression of the polynomials P^+ and P^- and/or symbolic computations, the values that the polynomials take, are calculated directly from the data flow graph. In order to do this, a new arithmetic is defined (Fig. 3).

The idea is to replace the classical single-value operands by two-value operands. The upper value (v^+) represents the value of the polynomial P⁺. The lower value (v^-) represents the value of the polynomial P⁻. Fig. 3 shows the redefinition of the operations used (*, +, -).



Fig. 3. Modified interval arithmetic definition.

oruci	oruer	Result order	
n	m	max(n, m)	
n	m	max(n, m)	
n	m	n + m	
	n n n	order order n m n m n m n m	

Table II. Input order calculation

In addition to the polynomial values, the polynomial order is also needed to calculate the bound functions. This order is also calculated with an operation redefinition, in a similar way to polynomial values. Table II presents the new operation definitions.

E. Extension to N dimensions

The basic idea of the extension to N dimensions is to split the problem into N sub-problems and apply the one-dimension approach (commented in subsection III.A) to each subproblem.

All these subspaces will have a common edge: the straight line (diagonal line) that joines the $X_I \equiv (x_{I1}, ..., x_{IN})$ and $X_S \equiv (x_{S1}, ..., x_{SN})$ extremities.

The diagonal line plays an important role in our technique and, in order to identify it easily, a variable change is performed. The elements of the new input vector $T \equiv (t_1, ..., t_N)$ verify the following properties:

$$0 \le t_1 \le 1, ..., 0 \le t_N \le 1$$

$$x_1 = (x_{S1} - x_{I1})^* t_1 + x_{I1}$$

...

$$x_N = (x_{SN} - x_{IN})^* t_N + x_{IN}$$
(11)

It is very easy to characterize the diagonal points with the new vector:

Diagonal Point = (t, t, ..., t, t) $0 \le t \le 1$

Additionally, the definition of the subspaces is also very simple. For example, the subspace 'i' is defined as the set of points $(t_1,...,t_i,...,t_N)$ in which the value of the 'i' dimension coordinate is less than the other values, that is, the set that verifies that $t_i \leq t_i \quad \forall j \neq i$, $j \in [1,N]$ and $ti \in [0,1]$.

In order to calculate the bound of a subspace, the algorithm makes use of the increasing and decreasing character of the polynomials P⁺ and P⁻. Let $T \equiv (t_1, ..., t_N)$ be an N dimensional input vector and $\Delta T \equiv (\Delta t_1, ..., \Delta t_N)$ be a positive shift ($\Delta t_1 \ge 0$, ..., $\Delta t_N \ge 0$). Thus, the positive (increasing) and negative (decreasing) polynomials verify:

$$P^{+}(T) \le P^{+}(T + \Delta T)$$

$$P^{-}(T) \ge P^{-}(T + \Delta T)$$
(12)

Let us focus on the polynomial P^+ (the discussion for the polynomial P^- is similar). As a consequence of the previously commented property, in the subspace 'i', the value of the polynomial at a point, T, verifies that:

Where:

$$T = (t_1, ..., t_i, ..., t_N), t_i \le t_j \quad \forall \ j \ne i$$

$$T_D = (t_i, ..., t_i, ..., t_i), Diagonal \ point$$

$$T_E = (1, ..., t_i, ..., 1), Extreme \ point$$

 $P^+(T_D) \le P^+(T) \le P^+(T_F)$

Expression (13) allows transforming the N-dimension problem into a 1-dimensional problem. The basic idea is that T_D and T_E are functions of only one variable, $t_i = t$. Thus, we could re-write (13) as:

Where:

$$D(\mathbf{v}) = \mathbf{v} \cdot (\mathbf{v}) = \mathbf{v} \cdot \mathbf{v}$$

$$T \equiv (t_1, \dots, t, \dots, t_N), t \le t_i \quad \forall \ j \ne i$$

 $P_{D}^{+}(t) < P^{+}(T) < P_{T}^{+}(t)$

 P_{D}^{+} and P_{E}^{+} are 1-dimensional polynomials and they can be bounded with the section III approach (equation (2)). Thus, equation (14) can be re-written and completed with equation (2) as:

$$P_{M}^{+}(t) \leq P_{D}^{+}(t) \leq P^{+}(T) \leq P_{E}^{+}(t) \leq P_{R}^{+}(t)$$

$$P_{M}^{-}(t) \geq P_{D}^{-}(t) \geq P^{-}(T) \geq P_{E}^{-}(t) \geq P_{R}^{-}(t)$$

$$P(T) = P^{+}(T) + P^{-}(T)$$

$$P_{R}^{+}(t) + P_{M}^{-}(t) \geq P(T) \geq P_{M}^{+}(t) + P_{R}^{-}(t)$$
(15)

(13)

(14)

Where:

$$T = (t_1, \dots, t, \dots, t_N), t \le t_j, \forall j \ne i$$
$$P_R(t) = mt + c$$
$$P_M(t) = \alpha t^n + \beta$$

In order to calculate the coefficient of the polynomials P_R^+ and P_M^+ for the 'i' subspace, the value of polynomial P⁺ has to be evaluated at 3 points: $X_I = (0,...,0)$, $X_S = (1,...,1)$ and $X_i = (1,...,1,0,1,...1)$. All the subspaces need the value of the X_I and X_S points, thus, only N+2 points will be needed to bound P⁺ in the N subspaces.

It is very important to highlight that every evaluated point is shared by 2^{N} N-dimensional intervals, thus the average number of points evaluated per interval is close to 1 in our verification algorithm.

Using equations (11), the MODIA algorithm estimates the P(T) bounds in every subspace. After the N subspaces have been bounded, the algorithm determines the bounds of P(T) in the complete input space.

IV. SYSTEM-LEVEL VERIFICATION BASED ON INTERVALS

The verification process begins from the behavioral description of the system. This specification is compiled in a control and data flow graph (CDFG) that contains only arithmetic (+, -, *) and relational operations, assertions and conditional statements ('if' constructions). At behavioral level, scheduling has not been performed, thus there is no state information in the graph.

For every path, there will be 'm' conditions that have to be satisfied and 'n' assertions that have to be checked. These constraints are modeled as the polynomial Ci (condition i) and Aj (assertion j), thus:

$$C_i(X) > 0 \qquad 1 \le i \le m \tag{16}$$

$$A_j(X) > 0 \qquad 1 \le j \le n \tag{17}$$

As we can see, all the constraints are converted into inequalities of the type "greater than".

The main goal of the verification algorithm is to find an input functional vector, $X \in [X_I, X_S]$, which satisfies the 'm' (16) inequalities and violates at least one of the (17) assertions. A point U violates the assertions polynomial $A_j(X)$ if $A_j(U) > 0$. In order to calculate the bounds of the polynomials C_i and A_i , the MODIA technique is used.

Additionally, during interval analysis the MODIA technique evaluates every polynomial in N+2 points. Let X_k ($1 \le k \le N+2$) be the set of points that MODIA has evaluated. When MODIA bounds a path for a particular input interval and the interval and path have not been previously removed, a heuristic cost function (18) is defined:

$$Cost = 0.5 \times R + 0.5 \times P \tag{18}$$

Parameter 'R' measures the probability that the conditional constraints are verified. Parameter 'P' measures the probability that an assertion is violated.

The complete pseudo-code of the verification algorithm is presented in the next section.

A. Pseudo-code of the Algorithm

The main goal of the algorithm is to find an input vector that violates an assertion. In order to do this the algorithm executes the next main loop for every assertion/property in the description.

Enumerate all paths that reach the property Select the complete input space as input interval For every path

Evaluate the cost function (eq. 18) End for; Loop (while there is a path to explore) Select the path with maximum cost Call path_evaluation function End loop;

The path_evaluation function tries to find an input interval that violates the assertion/properties. Its pseudo-code is:

Initialize the interval list with the complete input interval *Loop(while there is an input interval in the interval list)* Select the input interval with maximum cost Split the selected interval into 2 subintervals Loop (for each subinterval) Loop (for each condition) Bound the polynomial with MODIA If (lower bound of the conditional > 0) then *Eliminate the condition Elseif(upper bound of the conditional* ≤ 0 *) then* Remove the input interval from the list Next subinterval iteration Else Calculate the parameter R Endif; End loop: *Loop(for every assertion of the path)* Bound the polynomial with MODIA If(a point violates one assertion) then Counterexample detected *Elseif(upper bound of the assertion* ≤ 0) *then* Remove the input interval from the list Next subinterval iteration Else Calculate the parameter P Endif; End loop; Evaluate the cost function of equation 18 Insert the new interval in the interval list End loop: End loop; A counterexample cannot be found.

V. EXPERIMENTAL RESULTS

In order to validate the proposed technique, four behavioral system descriptions have been proposed. The test benches are data-dominated and they include non-linear behaviors and from 8-bit to 36-bit wide data path operations. Table III shows some properties of the selected test bench.

The performance of the proposed methodology has been compared with two classical model checking tools (SPIN [4] and SMV[3]) and two commercial tools: a demo version of [6] (Commercial1) and a full version of [5] (Commercial2). The results are presented in the Table IV. The table shows the CPU time that the tools need to generate a correct answer. The term OFL (Out oF Limit) normally identifies situations in which the program was aborted because the computer does not have enough memory resources. The CPU times in Table IV correspond to seconds on a Pentium III with 256 MB of RAM at 300 Mhz under Windows 2000. In the case of the "Commercial2" tool, the CPU times correspond to Sun Fire V120 Ultra-Sparc IIi with 512 MB of RAM running at 550 Mhz.

The tool 'SPIN' is not able to automatically explore the input space, thus an exhaustive search procedure has been included in the system description. SMV and SPIN are able to check simple designs but they run "out of limits" (OFL) when the size of the input space grows. The "Commercial 1" tool (Verity-Check) is able to verify nearly all the proposed examples. Only in one case, the tool was not able to provide a solution after 12 hours of CPU time. The "Commercial 2" tool (FormalCheck) runs on a different platform (Sun Workstations) thus, its execution times are not totally equivalent to the other tools. This tool has problems with the "Space4" test bench because it is not able to handle relational operators with more than 30 bits.

The proposed methodology spends less than a second to verify the proposed test bench. These results are better than other tools and only SPIN is able to obtain similar results in two examples.

VI. CONCLUSIONS AND FUTURE WORK

A method to check properties (or assertions) in behaviorallevel system specifications has been presented. The method is based on a modified interval analysis (MODIA) that can be directly computed over the CDFG, requiring (on average) the computation of about one new point per interval, reducing the classical overestimation problem of the IA techniques. Moreover, it is independent of the coding style of the algorithm. Using MODIA, the proposed verification algorithm explores the input space, looking for input intervals in which the violation of the assertion is more probable. The algorithm can also be used to automatically generate functional vectors that exercise predefined paths or properties. These vectors could be used to increase functional coverage metrics or replace random test generators.

The proposed technique has been evaluated with different examples and it produces better results than commercial and classical research model checkers.

The future work in this line is focused on two areas. The goal of the first is to extend the proposed methodology to handle RT-level descriptions. The second is to try to improve the methodology in order to avoid unrolling the loops and support more operators, concurrency, temporal expressions, and modular descriptions.

Benchmark	#input vectors	#solutions	#constraints	Polynomial order
Simple	256 ²	2509	3	2
Conditional	256 ²	1	2	2
Space3	256^{3}	1	3	2
Space4	256^4	2	6	4

Table III. Test bench properties

	SPIN	SMV	Comm. 1	Comm.2	MODIA
Simple	1 s	84 s	9 s	7 s	1 s
Conditional	1 s	OFL	120 s	100 s	1 s
Space3	OFL	OFL	22796 s	1637 s	1 s
Space4	OFL	OFL	> 12 hours	OFL	1 s

Table IV: Comparison with property checkers.

REFERENCES

- D. Ziegenbein, F. Wolf, K. Richter, M. Jersak, R. Ernst, "Interval-Based Analysis of Software Processes", *LCTES '2001*, pages 94-101, Snowbird, Utah, USA, June 2001.
- [2] J.P. Marques-Silva, K. A. Sakallah, "Boolean Satisfiability in Electronic Design Automation", DAC 2000.
- [3] K.L. McMillan. "Symbolic Model Checking: An approach to the State Explosion Problem". Kluwer Academic. 1993.
- [4] G. Holzmann. "The Model Checker SPIN". IEEE Trans. On Software Engineering. Vol 23, No. 5. May 1997.
- [5] FormalCheck. Cadence Design System.
- http://www.cadence.com/datasheets/formalcheck.html [6] Verity-Check. http://www.veritable.com/
- Computer/DesignVerityCheck.pdf [7] A. Biere, A. Cimatti, E. M. Clarke, M
- [7] A. Biere, A. Cimatti, E. M. Clarke, M. Fujita, Y. Zhu, "Symbolic Model Checking Using SAT procedures instead of BDDs*". Proc. of DAC'99. 1999.
- [8] F. Fallah, S. Nevadas, K. Keutzer. "Functional Vector Generation for HDL models Using Linear Programming and Boolean Satisfiability". IEEE Trans. of Computer-Aided Design of Integrated Circuits and Systems. Vol 20. No 8. August 2001.
- [9] Z. Zeng, P. Kalla, M. Ciesielski. "LPSAT: A Unified Approach to RTL Satisfiability". Proc. of DATE'01. 2001.
- [10] C. Huang, K. Cheng, "Assertion Checking by Combined Word-level ATPG and Modular Arithmetic Constraint-Solving Techniques", DAC 2000.
- [11] R. E. Bryant, "Graph-based algorithms for Boolean function manipulation", IEEE Transactions on Computers 35(8): (677-691), 1986.
- [12] R.E. Moore. Interval analysis. Prentice-Hall, 1966.
- [13] D. Ziegenbein, F. Wolf, K. Richter, M. Jersak, R. Ernst, "Interval-Based Analysis of Software Processes", LCTES '2001, Snowbird, Utah, USA, June 2001.
- [14] I. Ugarte, P. Sanchez, "System Verification Based on Modified Interval Analysis", ETW'03. 2003.
- [15] A. Biere, "SAT & ATPG in Formal verification". Embedded Tutorial in ICCAD'02. 2002.
- [16] P. Sanchez, S. Dey, "Simulation-based system-level verification using polynomials". HLDVT'99. 1999.